# Special requirements for ILC TPC Electronics

LCTPC meeting ALCPG007 Workshop Fermilab, Oct 21, 2007 *J.P. Martin, University of Montreal* 

# R&D program, history and future Detector

2007	<ul> <li>Prototype TPCs (Carleton-Victoria)</li> <li>Gas studies</li> <li>Pad design studies</li> <li>MPGDs</li> <li>Resistive film</li> </ul>	<ul> <li>Designed for R&amp;D requirements</li> <li>200 MHz VME FADCs</li> <li>Adapt legacy material (ALEPH,STA</li> <li>High density 48 channel cards (65 MHz VF48)</li> </ul>	AR)
2010	« LC TPC » -Anode pad modules Gem + small pad MicroMgas + resistive film	Objective: Produce a Realistic model for electronics using existin ASICS and off the shelf components - Optimize for detector requirement - Address heat management	Ig S
	Final detector	<b>Objective: Integrated electroni</b> - ASIC	CS
2017	2 J.P.	Martin, University of Montreal, LCTPC 22/10	0/2007

#### Typical tracking detector readout elements



#### Preamplifier requirements

The density of collected electrons is inhomogeneous (effect of the finite number of primary clusters) -The time structure is different in adjacent pads for the same track - There is an statistical error dependent on 1/SQRT(Number of clusters) over the pad

- There is an additional uncorrelated error due to the electronic noise Account for every cluster without bias, irrespective of their time of arrival Make the charge evaluation independernt of the time structure

The GEMs or MicroMegas are operated at low gain to reduce the positive ion backflow. Low noise preamplifiers are mandatory

### Time structure of the pad signals in a large TPC

- Case 1: Track almost parallel to the pad plane => The electrons drift for a long distance (meters) => Longitudinal diffusion (200 – 500 ns, gaussian)
- Case 2: The track passes close to the pad =>
   It has a large θ angle =>

Spread in arrival times (>500 ns, uniform)



• All cases: Transit time of electrons in the induction gap : adds <u>50-100 ns for GEM</u>)

Bottom line: - The charge collection process at the pads is <u>long</u> - It <u>varies</u> significantly from track to track

#### $\bullet \bullet \bullet$ Charge collection time structure artefacts

Exemple: collection of 3 unit charge clusters with a different time structure on 2 adjacent pads



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# Typical digital signal processing

« Moving window deconvolution » FIR transform:

$$N_{k} = A_{k} - A_{(k-w)} + (1/TauPreamp) \sum_{i=1,w}^{k} A_{(k-i)}$$

Conditions: Window width (w) = 400 ns Preamp decay = 1000 ns

#### « Boxcar » low-pass filter, T samples:

$$M_{k} = (\sum_{i=0,T-1} N_{(k-i)}) / T$$

Conditions: T = 75



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#### Note: Fast timing is useful for the timing or trigger channel

## Carleton – Montreal short term R&D plans

Prototype card built and tested in 2007:

High density card built with off the shelf components FPGA based: platform to test firmware signal processing algorithms

2008:

Include the new preamp ASIC Include compatibility with power pulsing

# Prototype card built and tested in 2007

- o 16 channels ( 2 octal 50 MS/sec FADCs with LVDS serial outputs)
- o Cyclone III FPGA (Low cost, low power)
- USB 2 interface (for test systems) Presently:
- No power pulsing
- o Preamp not on board

2008 => to be transformed into a compact 128 channel card with preamp ASIC and power pulsing capability

#### Develop electronics compatible with power pulsing

- o Based on the KOPIO design (OTS components)
- o Use latest serial output (LVDS) FADCs and FPGA
- Total area of FADCs + FPGA smaller than the altro chip
- o Design to match the TPC pad pitch
- o Include power pulsing capability
- Validate on existing test TPCs
- o BUT: Not cost effective for large production
  - (> 100K channels)
  - (Will need ASIC in due time)

# Power pulsing principle

- Power supplies designed for average power rather than peak power
- Distributed charge storage capacitors close to the switchable loads supply the peak current. Allow ~ 1 volt drop during power on cycle
- Local « Low dropout » voltage regulators keep voltage stable on the electronics
- Properly controlled constant current sources recharge the capacitors between cycles